# MJ10006 MJ10007

## Designers Data Sheet

#### SWITCHMODE SERIES NPN SILICON POWER DARLINGTON TRANSISTORS WITH BASE-EMITTER SPEEDUP DIODE

The MJ10006 and MJ10007 Darlington transistors are designed for high-voltage, high-speed, power switching in inductive circuits where fall time is critical. They are particularly suited for line operated switchmode applications such as:

- Switching Regulators
- Inverters
- Solenoid and Relay Drivers
- Motor Controls
- Deflection Circuits

Fast Turn-Off Times

30 ns Inductive Fall Time  $-25^{\circ}\text{C}$  (Typ)

500 ns Inductive Storage Time  $-25^{\circ}$ C (Typ) )

Operating Temperature Range -65 to +200°C

100°C Performance Specified for:

Reversed Biased SOA with Inductive Loads

Switching Times with Inductive Loads

Saturation Voltages

Leakage Currents



#### MAXIMUM BATINGS

INAXINON RATINGS				
Rating	Symbol	MJ10006	MJ10007	Unit
Collector-Emitter Voltage	V <sub>CEO</sub>	350	400	Vdc
Collector-Emitter Voltage	VCEX	400	450	Vdc
Collector-Emitter Voltage	VCEV	450	500	Vdc
Emitter Base Voltage	VEB	8.0		Vdc
Collector Current — Continuous — Peak (1)	IC ICM	10 20		Adc
Base Current — Continuous — Peak (1)	I <sub>B</sub>	2.5 5.0		Adc
Total Power Dissipation @ T <sub>C</sub> = 25°C @ T <sub>C</sub> = 100°C Derate above 25°C	PD	150 100 0.86		Watts W/OC
Operating and Storage Junction Temperature Range	T <sub>J</sub> ,T <sub>stg</sub>	-65 to +200		°C

#### THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	R <sub>0</sub> JC	: 1.17	°C/W
Maximum Lead Temperature for Soldering Purposes: 1/8" from Case for 5 Seconds	TL	275	°C

(1) Pulse Test: Pulse Width = 5.0 ms, Duty Cycle ≤ 10%.

#### 10 AMPERE

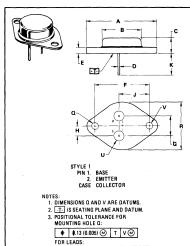
**NPN SILICON** 

### POWER DARLINGTON **TRANSISTORS**

350 AND 400 VOLTS 150 WATTS

#### Designer's Data for "Worst Case" Conditions

The Designers Data Sheet permits the design of most circuits entirely from the information presented. Limit data - representing device characteristics boundaries are given to facilitate "worst case" design.



	MILLIMETERS		INCHES		
DIM	MIN	MAX	MIN	MAX	
A	-	39.37	-	1.550	
В	_	21.08	-	0.830	
C	6.35	7,62	0.250	0,300	
D	0.97	1.09	0.038	0.043	
E	1.40	1.78	0.055	0.070	
F	30,15	BSC	1.187	7 BSC	
G	10.92 BSC		0.430 BSC		
н	5.46 BSC		0.215 BSC		
7	16,89 BSC		0.665 BSC		
ĸ	11.18	12.19	0.440	0.480	
┓	3,81	4.19	0.150	0.165	
R	-	26,67	-	1,050	
U	4.83	5.33	0.190	0.210	
v	3.81	4.19	0.150	0.165	

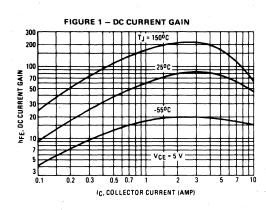
**CASE 1-05** TO-204AA

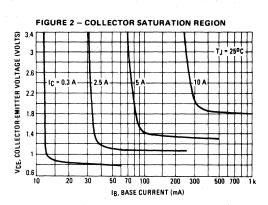
## MJ10006, MJ10007

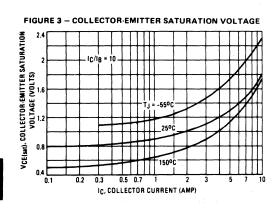
Characteristic	Symbol	Min	Тур	Max	Unit
OFF CHARACTERISTICS					
Collector-Emitter Sustaining Voltage (Table 1)	V <sub>CEO(sus)</sub>				Vdc
(IC = 250 mA, IB = 0, V <sub>clamp</sub> = Rated V <sub>CEO</sub> ) MJ10006		350	-	-	1
MJ10007		400	_		
Collector-Emitter Sustaining Voltage (Table 1, Figure 12)	VCEX(sus)				Vdc
(I <sub>C</sub> = 1 A, V <sub>clamp</sub> = Rated V <sub>CEX</sub> , T <sub>C</sub> = 100°C) MJ10006	1 1	400	-	-	
MJ10007		450		-	
(I <sub>C</sub> = 5 A, V <sub>clamp</sub> = Rated V <sub>CEX</sub> , T <sub>C</sub> = 100°C) MJ10006	}	275		_	
MJ10007	l+	325	<del></del>		- 0 -
Collector Cutoff Current	CEV			0.25	mAdc
(V <sub>CEV</sub> = Rated Value, V <sub>BE(off)</sub> = 1.5 Vdc)	1	_		5.0	
(V <sub>CEV</sub> = Rated Value, V <sub>BE(off)</sub> = 1.5 Vdc, T <sub>C</sub> = 150°C)	1		<del> </del>	5.0	mAdc
Collector Cutoff Current (VCE = Rated VCEV, RBE = 50 $\Omega$ , TC = 100°C)	ICER	-		3.0	
Emitter Cutoff Current	IEBO		<del> </del>	175	mAdc
(V <sub>EB</sub> = 2 Vdc, I <sub>C</sub> = 0)	, EBO				
SECOND BREAKDOWN	L		1		L
Second Breakdown Collector Current with base forward biased	le"		See Figure 1	 1	
	IS/b		See rigure r	<u>'</u>	L
ON CHARACTERISTICS (2)	<del></del>		,		,
DC Current Gain	hFE	40	1	500	_
(I <sub>C</sub> = 2.5 Adc, V <sub>CE</sub> = 5.0 Vdc) (I <sub>C</sub> = 5.0 Adc, V <sub>CE</sub> = 5.0 Vdc)		30		300	
Collector-Emitter Saturation Voltage	V <sub>CE(sat)</sub>		<del> </del>		Vdc
(I <sub>C</sub> = 5.0 Adc, I <sub>B</sub> = 250 mAdc)	(CE(sat)	_	_	1.9	""
(I <sub>C</sub> = 10 Adc, I <sub>B</sub> = 1.0 Adc)	}	_	_	2.9	1
(I <sub>C</sub> = 5.0 Adc, I <sub>B</sub> = 250 mAdc, T <sub>C</sub> = 100°C)	1	-	_	2.0	}
Base-Emitter Saturation Voltage	VBE(sat)		<u> </u>		Vdc
(I <sub>C</sub> = 5.0 Adc, I <sub>B</sub> = 250 mAdc)	00000	~	-	2.5	[
$(I_C = 5.0 \text{ Adc}, I_B = 250 \text{ mAdc}, T_C = 100^{\circ}\text{C})$	1	~	-	2.5	
Diode Forward Voltage (1)	Vf	-	3.0	5	Vdc
(I <sub>F</sub> = 5.0 Adc)	· 1				
DYNAMIC CHARACTERISTICS					
Small-Signal Current Gain	Ihfel	10	_	-	-
(I <sub>C</sub> = 1.0 Adc, V <sub>CE</sub> = 10 Vdc, f <sub>test</sub> = 1.0 MHz)					
Output Capacitance	Cob	60	_	275	ρF
$(V_{CB} = 10 \text{ Vdc}, I_E = 0, f_{test} = 100 \text{ kHz})$					<u> </u>
SWITCHING CHARACTERISTICS					
Resistive Load (Table 1)					
Delay Time (V <sub>CC</sub> = 250 Vdc, I <sub>C</sub> = 5.0 A,	td		0.05	0.2	μs
Rise Time $V_{CC} = 250 \text{ Vdc}, I_{C} = 5.0 \text{ A},$ $I_{B1} = 250 \text{ mA}, V_{BE(off)} = 5.0 \text{ Vdc}, t_{D} = 50 \mu s_{T}$	t <sub>r</sub>	-	0.25	0.6	μς
Storage Time   Duty Cycle ≤ 2.0%).	ts	_	0.5	1.5	μς
Fall Time	tf		0.06	0.5	μs
Inductive Load, Clamped (Table 1)					
Storage Time (IC = 5.0 A(pk), V <sub>clamp</sub> = Rated V <sub>CEX</sub> , I <sub>B1</sub> = 250 mA,	t <sub>sv</sub>	_	0.8	2.0	μѕ
Crossover Time VBE(off) = 5.0 Vdc, T <sub>C</sub> = 100°C)	tc		0.6	1.5	μς
Storage Time (IC = 5.0 A(pk), V <sub>clamp</sub> = Rated V <sub>CEX</sub> , I <sub>B1</sub> = 250 mA,	t <sub>sv</sub>		0.5	_	μs
Crossover Time VBE(off) = 5.0 Vdc, TC = 25°C)	t <sub>c</sub>		0.3	<del></del>	us us

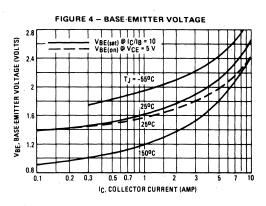
 <sup>(1)</sup> The internal Collector-to-Emitter diode can eliminate the need for an external diode to clamp inductive loads. Tests have shown that the Forward Recovery Voltage (V<sub>f</sub>) of this diode is comparable to that of typical fast recovery rectifiers.
 (2) Pulse Test: PW = 300 µs, Duty Cycle ≤ 2%.

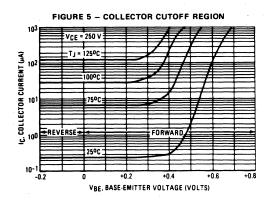
#### TYPICAL CHARACTERISTICS

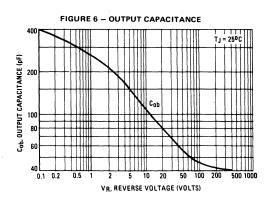












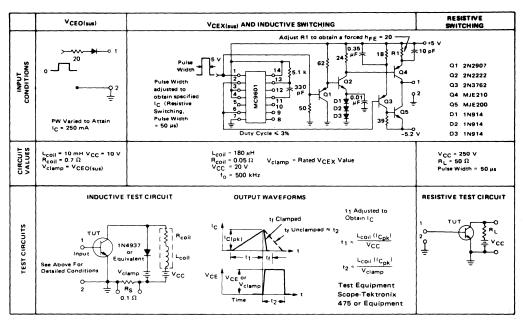
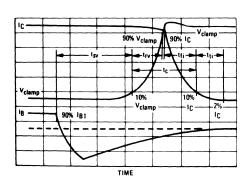


TABLE 1 - TEST CONDITIONS FOR DYNAMIC PERFORMANCE

#### FIGURE 7 - INDUCTIVE SWITCHING MEASUREMENTS



#### **SWITCHING TIMES NOTE**

In resistive switching circuits, rise, fall, and storage times have been defined and apply to both current and voltage waveforms since they are in phase. However, for inductive loads which are common to SWITCHMODE power supplies and hammer drivers, current and voltage waveforms are not in phase. Therefore, separate measurements must be made on each waveform to determine the total switching time. For this reason, the following new terms have been defined.

 $t_{sv}$  = Voltage Storage Time, 90% Ig1 to 10% V<sub>clamp</sub>  $t_{rv}$  = Voltage Rise Time, 10–90% V<sub>clamp</sub>  $t_{fi}$  = Current Fall Time, 90–10% I<sub>C</sub>

t<sub>ti</sub> = Current Tail, 10–2% I<sub>C</sub> t<sub>c</sub> = Crossover Time, 10% V<sub>clamp</sub> to 10% I<sub>C</sub>

An enlarged portion of the turn-off waveforms is shown in Figure  $\overline{7}$  to aid in the visual identity of these terms.

#### **SWITCHING TIME NOTES (continued)**

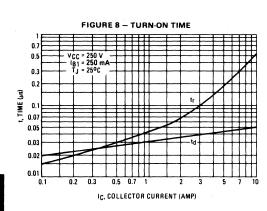
For the designer, there is minimal switching loss during storage time and the predominant switching power losses occur during the crossover interval and can be obtained using the standard equation from AN-222:

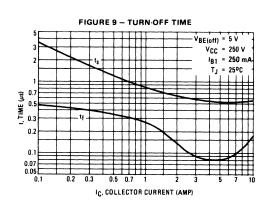
 $P_{SWT} = 1/2 V_{CCIC}(t_c) f$ 

In general,  $t_{rv}$  +  $t_{fi}$   $\simeq$   $t_{c}.$  However, at lower test currents this relationship may not be valid.

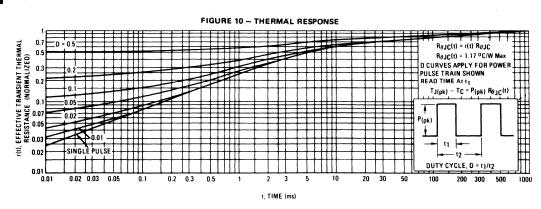
As is common with most switching transistors, resistive switching is specified at  $25^{\rm O}{\rm C}$  and has become a benchmark for designers. However, for designers of high frequency converter circuits, the user oriented specifications which make this a "SWITCHMODE" transistor are the inductive switching speeds (tc and tsv) which are guaranteed at  $100^{\rm O}{\rm C}$ .

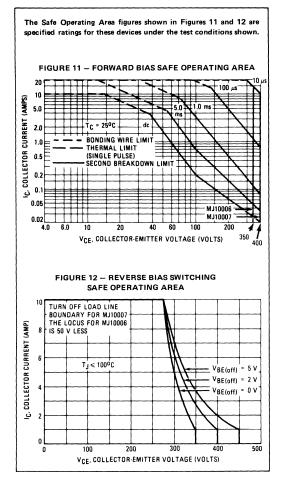
#### RESISTIVE SWITCHING PERFORMANCE











#### SAFE OPERATING AREA INFORMATION

#### **FORWARD BIAS**

There are two limitations on the power handling ability of a transistor: average junction temperature and second breakdown. Safe operating area curves indicate IC-VCE limits of the transistor that must be observed for reliable operation; i.e., the transistor must not be subjected to greater dissipation than the curves indicate.

The data of Figure 11 is based on  $T_C = 25^{\circ}C$ ;  $T_{J(pk)}$  is variable depending on power level. Second breakdown pulse limits are valid for duty cycles to 10% but must be derated when  $T_C \ge 25^{\circ}C$ . Second breakdown limitations do not derate the same as thermal limitations. Allowable current at the voltages shown on Figure 11 may be found at any case temperature by using the appropriate curve on Figure 13

TJ(pk) may be calculated from the data in Figure 10. At high case temperatures, thermal limitations will reduce the power that can be handled to values less than the limitations imposed by second breakdown.

### REVERSE BIAS

For inductive loads, high voltage and high current must be sustained simultaneously during turn-off, in most cases, with the base to emitter junction reverse biased. Under these conditions the collector voltage must be held to a safe level at or below a specific value of collector current. This can be accomplished by several means such as active clamping, RC snubbing, load line shaping, etc. The safe level for these devices is specified as  $V_{CEX(sus)}$  at a given collector current and represents a voltage-current condition that can be sustained during reverse biased turn-off. This rating is verified under clamped conditions so that the device is never subjected to an avalanche mode. Figure 12 gives the complete reverse bias safe operating area characteristics.

