## LINEAR INTEGRATED CIRCUITS

#### ADVANCE DATA

#### 2.5A POWER SWITCHING REGULATOR

- 5V TO 40V OUTPUT
- 2.5A OUTPUT CURRENT
- UP TO 100W OUTPUT POWER
- SOFT START
- PRECISE (± 4%) ON-CHIP REFERENCE
- VERY FEW EXTERNAL COMPONENTS
- 0 TO 100% DUTY CYCLE
- VERY HIGH EFFICIENCY (UP TO 90%)
- THERMAL SHUTDOWN

The L4960 is a monolithic power switching regulator delivering 2.5A at a voltage variable from 5V to 40V in step down configuration.

Features of the device include current limiting, soft start, thermal protection and 0 to 100% duty cycle for continuous operation mode.

The L4960 is mounted in a Heptawatt plastic power package and requires very few external components.

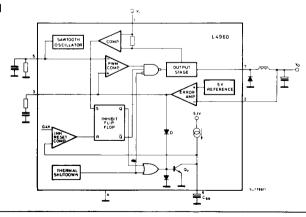
Efficient operation at switching frequencies up to 200KHz allows a reduction in the size and cost of external filter components.



#### ABSOLUTE MAXIMUM RATINGS

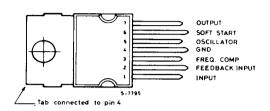
$V_{i}$	Input voltage	50	V
$V_i - V_7$	Input to output voltage difference	50	V
$V_7$	Negative output DC voltage	-1	V
	Negative output peak voltage (repetitive, tp ≤ 100ns)	17	V
$V_3, V_6$	Voltage at pin 3 and 6	5.5	V
V <sub>2</sub>	Voltage at pin 2	7	V
l <sub>3</sub>	Pin 3 sink current	1	mΑ
l <sub>5</sub>	Pin 5 source current	20	mΑ
$P_{tot}$	Power dissipation at $T_{case} \leq 90^{\circ}C$	15	W
$T_{j}, T_{stg}$	Junction and storage temperature	-40 to 150	°C

### **BLOCK DIAGRAM**



3/85

## CONNECTION DIAGRAM



## THERMAL DATA

R <sub>th j-case</sub> R <sub>th j-amb</sub>	Thermal resistance junction-case Thermal resistance junction-ambient	max	4 50	°C/W °C/W
Nth j-amb	i nermai resistance junction-ambient	max	50	C/W

## PIN FUNCTIONS

N° NAME		FUNCTION		
1	SUPPLY VOLTAGE	Unregulated voltage input. An internal regulator powers the internal logic.		
2	FEEDBACK INPUT	The feedback terminal of the regulation loop. The output is connected directly to this terminal for 5V operation; it is connected via a divider for higher voltages.		
3	FREQUENCY COMPENSATION	A series RC network connected between this terminal and ground determines the regulation loop gain characteristics.		
4	GROUND	Common ground terminal.		
5	OSCILLATOR	A parallel RC network connected to this terminal determines the switching frequency.		
6	SOFT START	Soft start time constant. A capacitor is connected between this terminal and ground to define the soft start time constant. This capacitor also determines the average short circuit output current.		
7	OUTPUT	Regulator output.		

# **ELECTRICAL CHARACTERISTICS** (Refer to the test circuit, $T_{amb} = 25^{\circ}C$ , $V_{i} = 35V$ , unless otherwise specified)

	Parameter	Test C	Conditions	Min.	Тур.	Max.	Unit
DYNAN	MIC CHARACTERISTICS			•	•	•	
Vo	Output voltage range	V <sub>i</sub> = 46V	I <sub>0</sub> = 1A	V <sub>ref</sub>		40	V
Vi	Input voltage range	Vo = Vref to 36V	1 <sub>0</sub> = 2A	9		46	V
ΔVo	Line regulation	V <sub>i</sub> = 10V to 40V	V <sub>o</sub> = V <sub>ref</sub> I <sub>o</sub> = 1A		15	50	m∨
ΔVo	Load regulation	Vo = Vref	I <sub>O</sub> = 0.5A to 2A		15	45	m∨
V <sub>ref</sub>	Internal reference voltage (pin 2)	V <sub>i</sub> = 9V to 46V	I <sub>O</sub> = 1A	4.8		5.2	V
∆V <sub>ref</sub> ∆T	Average temperature coefficient of refer, voltage	$T_j = 0^{\circ} \text{C to } 125^{\circ} \text{C}$ $I_0 = 1 \text{A}$			0.4		mV/°C
Vd	Dropout voltage	I <sub>0</sub> = 2A			2	3	V
lom	Maximum operating load current	V <sub>i</sub> = 9V to 46V V <sub>o</sub> = V <sub>ref</sub> to 36V		2.5			А
17L	Current limiting threshold (pin 7)	V <sub>i</sub> = 9V to 46V V <sub>o</sub> = V <sub>ref</sub> to 40V		3		4.5	А
Ish	Input average current	V <sub>i</sub> = 46V; output	short-circuit		30	60	mA
η	Efficiency	f = 100KHz	Vo = Vref		75		%
		I <sub>0</sub> = 2A	V <sub>o</sub> = 12V		85		%
SVR	Supply voltage ripple rejection	ΔV <sub>i</sub> = 2V <sub>rms</sub> f <sub>ripple</sub> = 100Hz V <sub>o</sub> = V <sub>ref</sub>	I <sub>O</sub> = 1A	50	56		dВ
∆f ∆Vi	Voltage stability of switching frequency	V <sub>i</sub> = 9V to 46V			0.5 _		%
∆f ∆Tj	Temperature stability of switching frequency	$T_j = 0^{\circ}C$ to 125°C			1		%
f <sub>max</sub>	Maximum operating switching frequency	Vo = Vref	I <sub>0</sub> = 2A	120	200		KHz
T <sub>sd</sub>	Thermal shutdown junction temperature				150		°C

junction temperature

's chartes

## **ELECTRICAL CHARACTERISTICS** (continued)

	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
OC CHA	RACTERISTICS					
l <sub>1Q</sub>	Quiescent drain current	Power ON Power OFF V <sub>i</sub> = 46V		30 15	40 20	mA mA
-1 <sub>7</sub> L	Output leakage current	Power OFF V <sub>i</sub> = 46V			2	mA
OFT S	TART				_	
leso	Source current			130		μΑ
1651	Sink current			70		μА
RROR	AMPLIFIER					
Vзн	High level output voltage	V <sub>2</sub> = 4.7V I <sub>3</sub> = 100μA	3.5			V
V3L	Low level output voltage	V <sub>2</sub> = 5.3V 1 <sub>3</sub> = 100μA			0.5	٧
1351	Sink output current	V <sub>2</sub> = 5.3V	100	150		μА
-l3so	Source output current	V <sub>2</sub> = 4.7V	100	150		μΑ
12	Input bias current	V <sub>2</sub> = 5.2V		2		μΑ
G <sub>v</sub>	DC open loop gain	V <sub>3</sub> = 1V to 3V	46	55		dB
OSCILL	ATOR					
-15	Oscillator source current		5			mA

## CIRCUIT OPERATION (refer to the block diagram)

The L4960 is a monolithic stepdown switching regulator providing output voltages from 5V to 40V and delivering 2.5A.

The regulation loop consists of a sawtooth oscillator, error amplifier, comparator and the output stage. An error signal is produced by comparing the output voltage with a precise 5V on-chip reference (zener zap trimmed to ± 4%).

This error signal in then compared with the sawtooth signal to generate the fixed frequency pulse width modulated pulses which drive the output stage.

The gain and frequency stability of the loop can be adjusted by an external RC network connected to pin 3. Closing the loop directly gives an output voltage of 5V. Higher voltages are obtained by inserting a voltage divider.

Output overcurrents at switch on are prevented by the soft start function. The error amplifier output is initially clamped by the external capaan output voltage of 5V. Higher voltages are citor  $C_{ss}$  and allowed to rise, linearly, as this capacitor is charged by a constant current source. Output overload protection is provided in the form of a current limiter. The load current is sensed by an internal metal resistor connected to a comparator. When the lead current exceeds a preset threshold this comparator sets a flip flop which disables the output stage and discharges the soft start capacitor. A second comparator resets the flip flop when the voltage across the soft start capacitor has fallen to 0.4V.

The output stage is thus re-enabled and the output voltage rises under control of the soft start network. If the overload condition is still present the limiter will trigger again when the threshold current is reached. The average short circuit current is limited to a safe value by the dead time introduced by the soft start network. The thermal overload circuit disables circuit operation when the junction temperature reaches about 150°C and has hysteresis to prevent unstable conditions.

Fig. 1 - Soft start waveforms

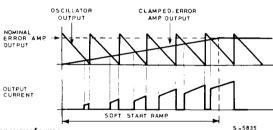


Fig. 2 - Current limiter waveforms

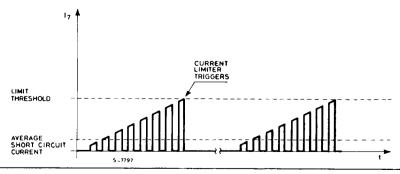


Fig. 3 - Test and application circuit

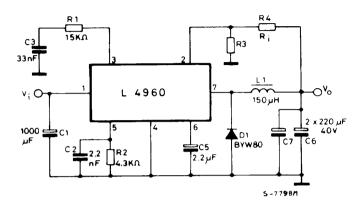
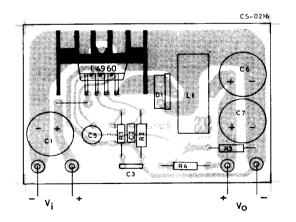


Fig. 4 - P.C. board and component layout of the circuit of fig. 3 (1:1 scale)



#### APPLICATION INFORMATION

Fig. 5 - A minimal 5V fixed regulator; Very few components are required

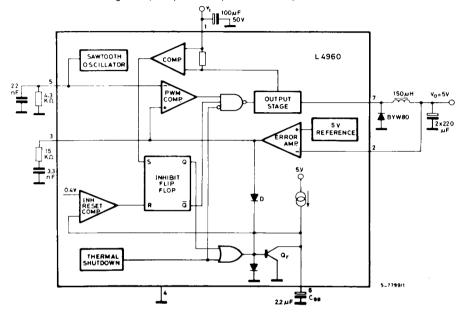
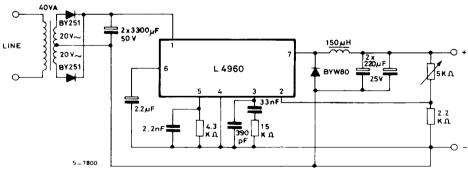


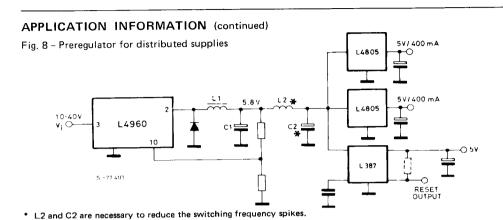
Fig. 6 - Programmable power supply



 $V_{o}$  = 5V to 15V  $I_{o}$  = 2.5A max load regulation (1A to 2A) = 10mV (V  $_{o}$  = 5V) line regulation (220V  $\pm$  15% and to  $I_{O}$  = 1A) = 15mV (V  $_{o}$  = 5V)

V ; = 35 V FUSE 2 x 3300µF 50 V E Y F SAWTOOH OSCILLATOR L 296 OUTPUT 5.1V REFERENCE O.2A 15KΩ 1000µF 25 V EKR ĸΩ THERMAL SHUTDOWN ---O RESET 2.2,uF F بر 2.2 INHIBIT 220µF SAWTOOH OSCILLATOR L 4960 OUTPUT STAGE 5 V REFERENCE −12V -○100mA 100 µF E**K**R 6.2 K Ω 5-7802/1

Fig. 7 - Microcomputer supply with + 5V, -5V, + 12V and -12V outputs



#### MOUNTING INSTRUCTION

The power dissipated in the circuit must be removed by adding an external heatsink. Thanks to the Heptawatt package attaching the heatsink is very simple, a screw or a compressione spring (clip) being sufficient. Between the heatsink and the package it is better to insert a layer of silicon grease, to optimize the thermal contact, no electrical isolation is needed between the two surfaces.

MECHANICAL DATA (Dimensions in mm)

Fig. 9 - Mounting example

